

For negative feedback system

$$a=1: y(t+1) = 1 - y(t)$$

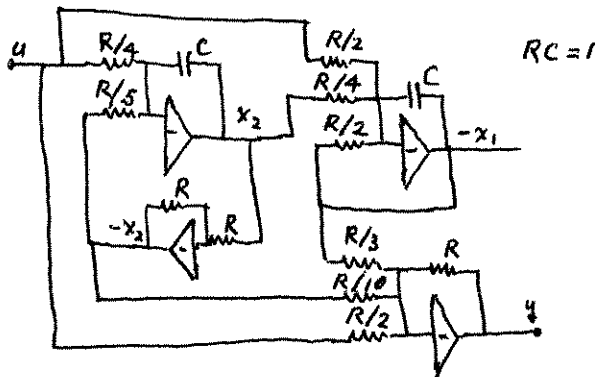
Because  $y(t) = 0$  for  $0 \leq t < 1$ , we have

$$\begin{aligned} y(t) &= 1 & \text{for } 1 \leq t < 2 \\ &= 0 & \text{for } 2 \leq t < 3 \\ &= 1 & \text{for } 3 \leq t < 4 \\ &\vdots \end{aligned}$$

$$a=0.5 \quad y(t+1) = 0.5(1 - y(t)) \quad \text{Then}$$

$$\begin{aligned} y(t) &= 0 & \text{for } 0 \leq t < 1 \\ &= 0.5 & \text{for } 1 \leq t < 2 \\ &= 0.5(1 - 0.5) = 0.25 & \text{for } 2 \leq t < 3 \\ &= 0.5(1 - 0.25) = 0.375 & \text{for } 3 \leq t < 4 \\ &\vdots \end{aligned}$$

$$\begin{aligned} 2.14 \quad \dot{x}_1 &= -2x_1 + 4x_2 + 2u \\ -\dot{x}_2 &= -5x_2 + 4u \\ -\dot{y} &= -3x_1 - 10x_2 + 2u \end{aligned}$$



2.15 (a) Apply Newton's law in the tangential direction:

$$u \cos \theta - mg \sin \theta = ml \ddot{\theta}$$

Define  $x_1 = \theta$ ,  $x_2 = \dot{\theta}$ . Then

$$\begin{cases} \dot{x}_1 = x_2 \\ \dot{x}_2 = -\frac{g}{l} \sin x_1 - \frac{u}{ml} \cos x_1 \end{cases}$$

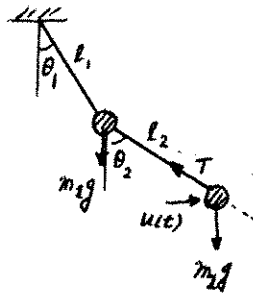
This is a nonlinear state equation.

If  $\theta$  is small, then  $\sin \theta \approx \theta$ ,  $\cos \theta \approx 1$ .

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -g/l & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ -1/ml \end{bmatrix} u$$

This is a linearized equation.

(b)



$$T = m_2 g \cos \theta_2 + u \sin \theta_2$$

$$u \cos \theta_2 - m_2 g \sin \theta_2 = m_2 l_2 \ddot{\theta}_2$$

$$\begin{aligned} T \sin(\theta_2 - \theta_1) - m_1 g \sin \theta_1 \\ = m_1 l_1 \ddot{\theta}_1 \end{aligned}$$

$$\begin{aligned} \text{Define } x_1 = \theta_1, \quad x_2 = \dot{\theta}_1 \\ x_3 = \theta_2, \quad x_4 = \dot{\theta}_2 \end{aligned}$$

Then we have

$$\dot{x}_1 = x_2$$

$$\begin{aligned} \dot{x}_2 = &-\frac{g}{l_1} \sin x_1 + \frac{m_2 g}{m_1 l_1} \cos x_3 \sin(x_3 - x_1) \\ &+ \frac{1}{m_1 l_1} \sin x_3 \sin(x_3 - x_1) u \end{aligned}$$

$$\dot{x}_3 = x_4$$

$$\dot{x}_4 = -\frac{g}{l_2} \sin x_3 + \frac{\cos x_3}{m_2 l_2} u$$

This is a nonlinear equation. If  $\theta_1 \approx 0$ , and  $\theta_2 \approx 0$ , then

$$\dot{x} = \begin{bmatrix} 0 & 1 & 0 & 0 \\ -g(m_1+m_2)/m_1 l_1 & 0 & m_2 g/m_1 l_1 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & -g/l_2 & 0 \end{bmatrix} x + \begin{bmatrix} 0 \\ 0 \\ 0 \\ 1/m_2 l_2 \end{bmatrix} u$$

This is a linearized equation.

$$2.16 \quad m \ddot{h} = f_1 - f_2 = k_1 \theta - k_2 u \quad (1)$$

$$I \ddot{\theta} + b \dot{\theta} = (l_1 + l_2) f_2 - l_1 f_1 \quad (2)$$

Define  $x_1 = h$ ,  $x_2 = \dot{h}$ ,  $x_3 = \theta$ ,  $x_4 = \dot{\theta}$ . Then

$$\begin{cases} \dot{x}_1 = x_2 \\ \dot{x}_2 = \frac{k_1}{m} x_3 - \frac{k_2}{m} u \\ \dot{x}_3 = x_4 \\ \dot{x}_4 = -\frac{l_1 k_1}{I} x_3 - \frac{b}{I} x_4 + (l_1 + l_2) k_2 u \end{cases}$$

This state equation describes the airplane.

Taking the Laplace transform of (1) and (2) and assuming  $I \approx 0$  yield

$$ms^2 \hat{h}(s) = k_1 \hat{\theta}(s) - k_2 \hat{u}(s)$$

$$bs \hat{\theta}(s) = (l_1 + l_2) k_2 \hat{u}(s) - k_1 l_1 \hat{\theta}(s)$$

$$\hat{\theta}(s) = \frac{(l_1 + l_2) k_2}{bs + k_1 l_1} \hat{u}(s)$$

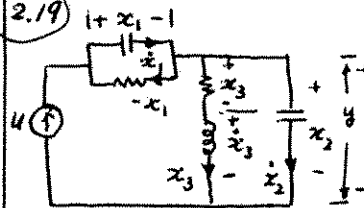
$$ms^2 \hat{h}(s) = \left( \frac{(l_1 + l_2) k_1 k_2}{bs + k_1 l_1} - k_2 \right) \hat{u}(s)$$

$$\hat{g}(s) = \frac{\hat{h}(s)}{\hat{u}(s)} = \frac{k_1 k_2 l_2 - k_2 bs}{ms^2 (bs + k_1 l_1)}$$

tank; therefore, we must compute  $\hat{g}(s) = \hat{y}(s)/\hat{u}(s)$  as a unit and do not have

$$\hat{g}(s) = \hat{g}_1(s) \hat{g}_2(s).$$

2.19



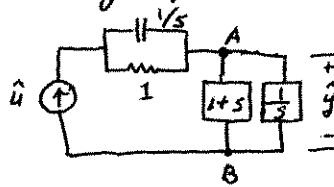
$$\dot{x}_1 = -x_1 + u$$

$$x_3 + \dot{x}_2 = u$$

$$x_3 + \dot{x}_3 = x_2 = y$$

$$\dot{x} = \begin{bmatrix} -1 & 0 & 0 \\ 0 & 0 & -1 \\ 0 & 1 & -1 \end{bmatrix} x + \begin{bmatrix} 1 \\ 1 \\ 0 \end{bmatrix} u \quad y = [0 \ 1 \ 0] x$$

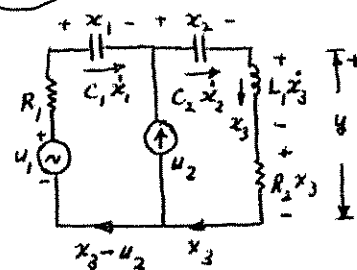
The transfer function can be computed from  $C(sI - A)^{-1}b + d$  or, more easily, using impedances as



$$Z_{AB} = \frac{\frac{1}{s}(s+1)}{\frac{1}{s} + s + 1} = \frac{s+1}{s^2 + s + 1}$$

Thus we have  $\hat{y}(s) = \frac{s+1}{s^2 + s + 1} \hat{u}(s)$ .

2.20



The voltage across  $R_1$  is  $R_1(x_3 - u_2)$ .

We have

$$C_1 \dot{x}_1 = x_3 - u_2$$

$$C_2 \dot{x}_2 = x_3$$

From the outer loop, we have

$$L_1 \dot{x}_3 = -x_1 - x_2 - (x_3 - u_2)R_1 - R_2 x_3 + u_1. \text{ Thus}$$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} 0 & 0 & 1/C_1 \\ 0 & 0 & 1/C_2 \\ -1/L_1 & -1/L_1 & -(R_1 + R_2)/L_1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} 0 & -1/C_1 \\ 0 & 0 \\ 1/L_1 & R_1/L_1 \end{bmatrix} \begin{bmatrix} u_1 \\ u_2 \end{bmatrix}$$

$$y = L_1 \dot{x}_3 + R_2 x_3 = -x_1 - x_2 - (x_3 - u_2)R_1 + u_1$$

$$= [-1 \ -1 \ -R_1] x + [1 \ R_1] \begin{bmatrix} u_1 \\ u_2 \end{bmatrix}$$

2.17  $m\ddot{y} = -kx - mg$

$$x_1 = y, \quad x_2 = \dot{y}, \quad x_3 = m, \quad u = m$$

$$\begin{cases} \dot{x}_1 = x_2 \\ \dot{x}_2 = \frac{-k}{m} x_1 - g \\ \dot{x}_3 = u \end{cases}$$

a nonlinear equation

2.18 Following Example 2.9, we have

$$y_1 = \frac{x_1}{R_1}, \quad A_1 dx_1 = (u - y_1) dt$$

$$\text{Thus } \begin{cases} \dot{x}_1 = \frac{-1}{A_1 R_1} x_1 + \frac{1}{A_1} u \\ y_1 = \frac{1}{R_1} x_1 \end{cases}$$

$$\hat{g}_1(s) = \frac{\hat{y}_1(s)}{\hat{u}(s)} = \frac{1}{R_1} \left( s + \frac{1}{A_1 R_1} \right)^{-1} \frac{1}{A_1} = \frac{1}{A_1 R_1 s + 1}$$

Similarly

$$\hat{g}_2(s) = \frac{\hat{y}_2(s)}{\hat{u}_2(s)} = \frac{1}{A_2 R_2 s + 1}$$

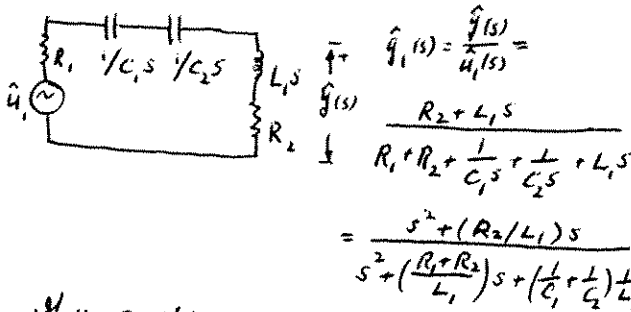
The transfer function from  $u$  to  $y$  is

$$\hat{g}(s) = \hat{g}_1(s) \hat{g}_2(s) = \frac{1}{(A_1 R_1 s + 1)(A_2 R_2 s + 1)}$$

The transfer function from  $u$  to  $y_1$  in

Fig. 2.13 depends on  $x_2$  of the second

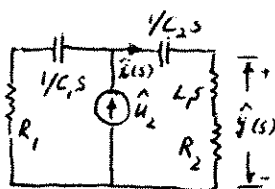
We use impedances to compute transfer functions, if  $u_2 = 0$ , the circuit reduces to



$$\hat{g}_1(s) = \frac{\hat{y}(s)}{\hat{u}_1(s)} = \frac{R_2 + L_1 s}{R_1 + R_2 + \frac{1}{C_1 s} + \frac{1}{C_2 s} + L_1 s}$$

$$= \frac{s^2 + (R_2/L_1)s}{s^2 + \left(\frac{R_1 + R_2}{L_1}\right)s + \left(\frac{1}{C_1} + \frac{1}{C_2}\right)\frac{1}{L_1}}$$

If  $u_1 = 0$ , then



$$\hat{g}_2(s) = \frac{(R_1 + \frac{1}{C_1 s})\hat{u}_2(s)}{R_1 + R_2 + \frac{1}{C_1 s} + \frac{1}{C_2 s} + L_1 s}$$

$$\hat{g}_2(s) = \frac{\hat{y}(s)}{\hat{u}_2(s)} = \frac{(R_1 s + \frac{1}{C_1})(s + \frac{R_2}{L_1})}{s^2 + \left(\frac{R_1 + R_2}{L_1}\right)s + \left(\frac{1}{C_1} + \frac{1}{C_2}\right)\frac{1}{L_1}}$$

Therefore

$$\hat{y}(s) = \hat{g}_1(s)\hat{u}_1(s) + \hat{g}_2(s)\hat{u}_2(s)$$

$$= \begin{bmatrix} \hat{g}_1(s) & \hat{g}_2(s) \end{bmatrix} \begin{bmatrix} \hat{u}_1(s) \\ \hat{u}_2(s) \end{bmatrix}$$

Note that the denominator of  $\hat{g}_i(s)$  is different from  $\det(sI - A)$ . The former has degree 2, the latter has degree 3.

2.21 Let  $I$  be the moment of inertia of the bar and mass about the hinge. Then

$$I\ddot{\theta} = l_2 u - k_1(\theta l_1)l_1 - k_2(l_2\theta - y)l_2$$

$$m_2 \ddot{y} = k_2(l_2\theta - y) \quad \text{for } \theta \text{ small.}$$

Define  $x_1 = \theta$ ,  $x_2 = \dot{\theta}$ ,  $y = x_3$  and  $\dot{x}_4 = \dot{y}$ . Then

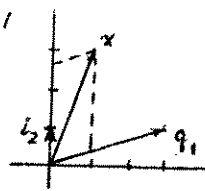
$$\dot{x} = \begin{bmatrix} 0 & 1 & 0 & 0 \\ -(k_1 l_1^2 + k_2 l_2^2)/I & 0 & k_2 l_2/I & 0 \\ 0 & 0 & 0 & 1 \\ k_2 l_2/m_2 & 0 & -k_2/m_2 & 0 \end{bmatrix} x + \begin{bmatrix} 0 \\ \frac{l_2}{I} u \\ 0 \\ 0 \end{bmatrix}$$

$$y = \begin{bmatrix} 0 & 0 & 1 & 0 \end{bmatrix} x$$

$$\hat{g}(s) = \frac{\hat{y}(s)}{\hat{u}(s)} = \frac{k_2 l_2^2}{I m s^4 + [m_2(k_1 l_1^2 + k_2 l_2^2) + I k_2] s^2 + k_1 k_2 l_1^2}$$

### Chapter 3

3.1

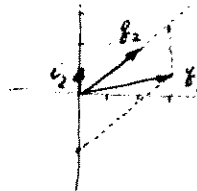


$$\text{Because } x = \frac{1}{3}q_1 + \left(2\frac{2}{3}\right)q_2$$

$$= \begin{bmatrix} q_1 & q_2 \end{bmatrix} \begin{bmatrix} 1/3 \\ 8/3 \end{bmatrix}$$

the representation of  $x$  with respect to  $\{q_1, q_2\}$  is  $\begin{bmatrix} 1/3 \\ 8/3 \end{bmatrix}$ . Indeed, we have

$$x = \begin{bmatrix} 3 & 0 \\ 1 & 1 \end{bmatrix} \begin{bmatrix} 1/3 \\ 8/3 \end{bmatrix} = \begin{bmatrix} 1 \\ 3 \end{bmatrix}$$



$$\text{Because } q_1 = -2e_2 + 1.5q_2$$

$$= \begin{bmatrix} e_2 & q_2 \end{bmatrix} \begin{bmatrix} -2 \\ 1.5 \end{bmatrix}$$

the representation of  $q_1$  with respect to  $\{e_2, q_2\}$  is  $\begin{bmatrix} -2 \\ 1.5 \end{bmatrix}$ . Indeed we have

$$q_1 = \begin{bmatrix} 0 & 2 \\ 1 & 2 \end{bmatrix} \begin{bmatrix} -2 \\ 1.5 \end{bmatrix} = \begin{bmatrix} 3 \\ 1 \end{bmatrix}$$

3.2

$$\|x_1\|_1 = 2 + 3 + 1 = 6$$

$$\|x_1\|_2 = \sqrt{4 + 9 + 1} = \sqrt{14} = 3.74$$

$$\|x_1\|_\infty = 3$$

$$\|x_2\|_1 = 1 + 1 + 1 = 3$$

$$\|x_2\|_2 = \sqrt{1 + 1 + 1} = \sqrt{3} = 1.732$$

$$\|x_2\|_\infty = 1$$

3.3

$$q_1 = x_1 / \|x_1\| = \frac{1}{3.74} \begin{bmatrix} 2 \\ -3 \\ 1 \end{bmatrix}$$

$$u_2 = x_2 - (q_1^T x_2) q_1 = \begin{bmatrix} 1 \\ 1 \\ 1 \end{bmatrix} - \frac{0}{3.74} q_1 = \begin{bmatrix} 1 \\ 1 \\ 1 \end{bmatrix}$$

$$q_2 = u_2 / \|u_2\| = \frac{1}{1.732} \begin{bmatrix} 1 \\ 1 \\ 1 \end{bmatrix}$$

Note that  $x_1$  and  $x_2$  are already orthogonal. Therefore  $q_1$  and  $q_2$  are their normalized vectors.

3.4 If  $n > m$ ,  $AA^T$  is symmetric and has rank  $m$ . If  $m = n$ , then  $A^T A = I_n$  and